EPC2071 – Enhancement Mode Power Transistor

V_{DS} , 100 V $R_{DS(on)}$, 1.7 m Ω typical, 2.2 m Ω max I_D, 64 A



Gallium Nitride's exceptionally high electron mobility and low temperature coefficient allows very low R_{DS(on)}, while its lateral device structure and majority carrier diode provide exceptionally low Q_G and zero Q_{RR}. The end result is a device that can handle tasks where very high switching frequency, and low on-time are beneficial as well as those where on-state losses dominate.

Application Notes:

- Easy-to-use and reliable gate
- Gate Drive ON = 5–5.25 V typical, OFF = 0 V (negative voltage not needed)
- Recommended dead time (half bridge circuit) ≤ 30 ns for best efficiency
- Top of FET (back side) is electrically connected to source

Maximum Ratings				
	PARAMETER	VALUE	UNIT	
V	Drain-to-Source Voltage (Continuous)	100	V	
V _{DS}	Drain-to-Source Voltage (\leq 10k pulses of \leq 5ms at 150°C)	120		
	Continuous (T _A = 25°C)	64	•	
I _D	Pulsed (25°C, T _{PULSE} = 300 μs)	350	A	
V	Gate-to-Source Voltage	6	V	
V _{GS}	Gate-to-Source Voltage	-4	V	
٦	Operating Temperature	-40 to 150	°C	
T _{STG}	Storage Temperature	-40 to 150		

Thermal Characteristics					
PARAMETER			UNIT		
R _{θJC}	Fhermal Resistance, Junction-to-Case (Case TOP)				
R _{θJB}	Thermal Resistance, Junction-to-Board (Case BOTTOM)		°C/W		
R _{0JA_JEDEC}	R _{0JA_JEDEC} Thermal Resistance, Junction-to-Ambient (using JEDEC 51-2 PCB)				
R _{0JA EVB}	Thermal Resistance, Junction-to-Ambient (using EPC90146 EVB)	31			

Die Size: 4.45 x 2.3 mm EPC2071 eGaN[®] FETs are supplied only in passivated die form with solder bars.

EFFICIENT POWER CONVERSION

Applications

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- 48 V DC-DC Converters
- BLDC Motor Drives
- Sync Rectification for AC/DC and DC-DC
- Point of Load Converters
- Solar Converters
- Lidar
- eMobility

Benefits

- Ultra High Efficiency
- No Reverse Recovery
- Ultra Low Q_G, Q_{GD}, Q_{OSS}
- Ultra Low R_{DS(on)}
- Ultra Small Footprint



PARAMETER		TEST CONDITIONS MIN		ТҮР	MAX	UNIT
BV _{DSS}	Drain-to-Source Voltage	$V_{GS} = 0 V, I_D = 0.15 mA$	100			V
I _{DSS}	Drain-Source Leakage	$V_{GS} = 0 V, V_{DS} = 80 V$		0.001	0.12	mA
I _{GSS}	Gate-to-Source Forward Leakage	$V_{GS} = 5 V$		0.03	3.2	
	Gate-to-Source Forward Leakage [#]	$V_{GS} = 5 V, T_J = 125 ^{\circ}C$		0.3	7.1	
	Gate-to-Source Reverse Leakage	$V_{GS} = -4 V$		0.006	0.17	
V _{GS(TH)}	Gate Threshold Voltage	$V_{DS} = V_{GS}, I_D = 13 \text{ mA}$	0.7	1.3	2.5	V
R _{DS(on)}	Drain-Source On Resistance	$V_{GS} = 5 \text{ V}, \text{ I}_{D} = 30 \text{ A}$		1.7	2.2	mΩ
V_{SD}	Source-Drain Forward Voltage [#]	$I_{S} = 0.5 \text{ A}, V_{GS} = 0 \text{ V}$		1.5		V

Defined by design. Not subject to production test

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Halogen-Free



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FORUM

Dynamic Characteristics [#] ($T_j = 25^{\circ}$ C unless otherwise stated)						
	PARAMETER	TEST CONDITIONS	MIN	ТҮР	MAX	UNIT
C _{ISS}	Input Capacitance			2664	3931	
C _{RSS}	Reverse Transfer Capacitance	$V_{DS} = 50 V, V_{GS} = 0 V$		5.4		
C _{OSS}	Output Capacitance			878	976	pF
C _{OSS(ER)}	Effective Output Capacitance, Energy Related (Note 1)	$V_{DS} = 0$ to 50 V, $V_{GS} = 0$ V		1058		
C _{OSS(TR)}	Effective Output Capacitance, Time Related (Note 2)			1422		
R _G	Gate Resistance			0.3		Ω
Q _G	Total Gate Charge	$V_{DS} = 50 \text{ V}, V_{GS} = 5 \text{ V}, I_D = 30 \text{ A}$		18	26	
Q _{GS}	Gate-to-Source Charge	$V_{DS} = 50 \text{ V}, \text{ I}_{D} = 30 \text{ A}$		6.0		
Q_{GD}	Gate-to-Drain Charge			1.8		
Q _{G(TH)}	Gate Charge at Threshold			4.5		nC
Q _{OSS}	Output Charge	$V_{DS} = 50 \text{ V}, V_{GS} = 0 \text{ V}$		71	82	
Q _{RR}	Source-Drain Recovery Charge (Note 3)			0		

Defined by design. Not subject to production test.

All measurements were done with substrate connected to source. Note 1: $C_{OSS(ER)}$ is a fixed capacitance that gives the same stored energy as C_{OSS} while V_{DS} is rising from 0 to 50% BV_{DSS}. Note 2: $C_{OSS(TR)}$ is a fixed capacitance that gives the same charging time as C_{OSS} while V_{DS} is rising from 0 to 50% BV_{DSS}. Note 3: GaN FET do not have an anti-parallel body diode, and hence do not exhibit reverse recovery. However they can operate in the

third quadrant, and their reverse conduction characteristic is shown in Figure 8.









Figure 2: Typical Transfer Characteristics





Figure 12: Transient Thermal Response Curves







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LAYOUT CONSIDERATIONS

GaN transistors generally behave like power MOSFETs, but at much higher switching speeds and power densities, therefore layout considerations are very important, and care must be taken to minimize layout parasitic inductances. The recommended design utilizes the first inner layer as a power loop return path. This return path is located directly beneath the top layer's power loop allowing for the smallest physical loop size. This method is also commonly referred to as flux cancellation. Variations of this concept can be implemented by placing the bus capacitors either next to the high side device, or next to the low side device, or between the low and high side devices, but in all cases the loop is closed using the first inner layer right beneath the devices.

A similar concept is also used for the gate loop, with the return gate loop located directly under the turn ON and OFF gate resistors.

Furthermore, to minimize the common source inductance between power and gate loops, the power and gate loops are laid out perpendicular to each other, and a via next to the source pad closest to the gate pad is used as Kelvin connection for the gate driver return path.

The EPC90146 Quick Start Guide – 100 V, 40 A Half-Bridge Development Board Using EPC2071 implements our recommended vertical inner layout.



Output: Set of the source o

Figure 13: Inner vertical layout for power and gate loops from EPC90146

Detailed recommendations on layout can be found on EPC's website: Optimizing PCB Layout with eGaN FETs.pdf

TYPICAL SWITCHING BEHAVIOR

The following typical switching waveforms are captured in these conditions:

- EPC90146 100 V, 40 A Half-bridge Development Board using EPC2071
- Gate driver: uP1966E with 0.4 $\Omega/0.7~\Omega$ pull-down/pull-up resistance
- External $R_G(ON) = 0 \Omega$, $R_G(OFF) = 0 \Omega$
- $V_{IN} = 64 \text{ V}, I_L = 32 \text{ A}$



TYPICAL THERMAL CONCEPT

The EPC2071 can take advantage of dual sided cooling to maximize its heat dissipation capabilities in high power density designs. Note that the top of EPC FETs are connected to source potential, so for half-bridge topologies the Thermal Interface Material (TIM) needs to provide electrical isolation to the heatsink.

Recommended best practice thermal solutions are covered in detail in How2AppNote012 - How to Get More Power Out of an eGaN Converter.pdf.



Figure 15: Exploded view of heatsink assembly using screws



Figure 16: A cross-section image of dual sided thermal solution

Note: Connecting the heatsink to ground is recommended and can significantly improve radiated EMI

The thermal design can be optimized by using the GaN FET Thermal Calculator on EPC's website.

EPC2071

TAPE AND REEL CONFIGURATION

8 mm pitch, 12 mm wide tape on 7" reel

Δ

DIE MARKINGS







Recommended stencil should be 4 mil (100 µm) thick, must be laser cut, openings per drawing.

The corner has a radius of R60 µm.

Intended for use with SAC305 Type 4 solder, reference 88.5% metals content.

EPC2071

MAX

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ADDITIONAL RESOURCES AVAILABLE

Solder mask defined pads are recommended for best reliability.



Figure 17: Solder mask defined versus non-solder mask defined pad



Figure 18: Effect of solder mask design on the solder ball symmetry

- Assembly resources https://epc-co.com/epc/Portals/0/epc/documents/product-training/Appnote_GaNassembly.pdf
- Library of Altium footprints for production FETs and ICs https://epc-co.com/epc/documents/altium-files/EPC%20Altium%20Library.zip
 (for preliminary device Altium footprints, contact EPC)

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